

# On the Completeness of the Lattice Factorization for Linear-Phase Perfect Reconstruction Filter Banks

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*Abstract*— In this letter, we re-examine the completeness of the lattice factorization for  $M$ -channel linear-phase perfect reconstruction filter bank (LPPRFB) with filters of the same length  $L = KM$  in [1]. We point out that the assertion of completeness in [1] is incorrect. Examples are presented to show that the proposed lattice structure in [1] is *not* complete when  $K > 2$ . In addition, we verify that the lattice structure in [1] is complete only when  $K \leq 2$ .

*Index Terms*— Lattice factorization, linear-phase perfect reconstruction filter bank, completeness.

## I. INTRODUCTION

Lattice factorization is one of the most attractive methods for the design and implementation of filter banks. An important concept associated with the lattice structure is *completeness*. Completeness of a lattice implies that the lattice can cover all possible solutions for any filter bank possessing certain desired properties, such as perfect reconstruction (PR), paraunitary (PU) and/or linear-phase perfect reconstruction (LPPR). In [1], a lattice structure for an  $M$ -channel linear-phase perfect reconstruction filter bank (LPPRFB) with all the filters of the same length  $L = KM$  was introduced. When  $M$  is even, the proposed lattice structure in [1] was asserted to be complete ([1], Theorem II), i.e., the lattice was supposed to cover all even-channel LPPRFBs with the same filter length. However, in this letter, we point out that the assertion of completeness in [1] is incorrect. Examples are presented to show that the proposed lattice structure in [1] is *not* complete when  $K > 2$ . In addition, we prove that the lattice structure in [1] is complete only when  $K \leq 2$ .

Hereafter, we use the following notations to denote certain special matrices. For a positive integer  $m$ ,  $\mathbf{I}_m$ ,  $\mathbf{J}_m$  and  $\mathbf{0}_m$  denote the  $m \times m$  identity matrix, reversal matrix and null matrix [1], respectively. Moreover,  $\mathbf{D}_{2m}$  is a  $2m \times 2m$  matrix as follows

$$\mathbf{D}_{2m} = \begin{bmatrix} \mathbf{I}_m & \mathbf{0}_m \\ \mathbf{0}_m & -\mathbf{I}_m \end{bmatrix}. \quad (1)$$

## II. LATTICE FACTORIZATION FOR LPPRFBs

Consider an  $M$ -channel ( $M$  even,  $M = 2m$ ) LPPRFB with all the analysis and synthesis filters of the same length  $L = KM$  each. Let  $\mathbf{E}(z)$  and  $\mathbf{R}(z)$  denote the corresponding analysis and synthesis polyphase matrix, respectively.

$\mathbf{E}(z)$  and  $\mathbf{R}(z)$  satisfy the LPPR condition as follows [1]

$$\mathbf{R}(z)\mathbf{E}(z) = z^{-l}\mathbf{I}_M, \quad l \geq 0, \quad (2)$$

$$\mathbf{E}(z) = z^{-(K-1)}\mathbf{D}_M\mathbf{E}(z^{-1})\mathbf{J}_M, \quad (3)$$

$$\mathbf{R}(z) = z^{-(K-1)}\mathbf{J}_M\mathbf{R}(z^{-1})\mathbf{D}_M. \quad (4)$$

Equation (2) is referred to as the PR property, and equations (3) and (4) are referred to as the LP property. Collectively, (2)-(4) are called the LPPR condition. For this subclass of LPPRFBs, the lattice factorization derived in [1] is as follows:

$$\mathbf{E}(z) = \mathbf{G}_{K-1}(z)\mathbf{G}_{K-2}(z)\cdots\mathbf{G}_2(z)\mathbf{G}_1(z)\mathbf{E}_0, \quad (5)$$

$$\mathbf{G}_i(z) = \frac{1}{2} \begin{bmatrix} \mathbf{U}_i & \mathbf{0} \\ \mathbf{0} & \mathbf{V}_i \end{bmatrix} \begin{bmatrix} \mathbf{I} & \mathbf{I} \\ \mathbf{I} & -\mathbf{I} \end{bmatrix} \begin{bmatrix} \mathbf{I} & \mathbf{0} \\ \mathbf{0} & z^{-1}\mathbf{I} \end{bmatrix} \begin{bmatrix} \mathbf{I} & \mathbf{I} \\ \mathbf{I} & -\mathbf{I} \end{bmatrix}, \quad (6)$$

$$\triangleq \frac{1}{2}\Phi_i\mathbf{W}\Lambda(z)\mathbf{W}$$

and

$$\mathbf{E}_0 = \frac{1}{\sqrt{2}} \begin{bmatrix} \mathbf{U}_0 & \mathbf{U}_0\mathbf{J} \\ \mathbf{V}_0\mathbf{J}_m & -\mathbf{V}_0 \end{bmatrix} = \frac{1}{\sqrt{2}} \begin{bmatrix} \mathbf{U}_0 & \mathbf{0} \\ \mathbf{0} & \mathbf{V}_0 \end{bmatrix} \begin{bmatrix} \mathbf{I} & \mathbf{J} \\ \mathbf{J} & -\mathbf{I} \end{bmatrix}, \quad (7)$$

where  $\mathbf{U}_i$ ,  $\mathbf{V}_i$  (for  $i = 0, 1, \dots, K-1$ ) are arbitrary  $m \times m$  invertible matrices. The synthesis polyphase matrix  $\mathbf{R}(z)$  can be obtained by inverting each analysis component one by one in  $\mathbf{E}(z)$  [1].

Although the above factorization can structurally enforce both LP and PR properties, it contains redundant free parameters. In [2], a simplified lattice factorization was presented, where all the matrices  $\mathbf{V}_i$  (for  $i = 1, \dots, K-1$ ) can be replaced by the identity matrix  $\mathbf{I}_m$  without loss of generality. That is, all  $\mathbf{G}_i(z)$  in (6) can be substituted by

$$\mathbf{G}_i(z) = \frac{1}{2} \begin{bmatrix} \mathbf{U}_i & \mathbf{0} \\ \mathbf{0} & \mathbf{I} \end{bmatrix} \begin{bmatrix} \mathbf{I} & \mathbf{I} \\ \mathbf{I} & -\mathbf{I} \end{bmatrix} \begin{bmatrix} \mathbf{I} & \mathbf{0} \\ \mathbf{0} & z^{-1}\mathbf{I} \end{bmatrix} \begin{bmatrix} \mathbf{I} & \mathbf{I} \\ \mathbf{I} & -\mathbf{I} \end{bmatrix}. \quad (8)$$

The simplified expression in (8) provides a much more efficient way for both design and implementation of LPPRFB. Later, we will use it to analyze the completeness of the lattice factorization.

## III. ON THE COMPLETENESS OF THE LATTICE FACTORIZATION

### A. Review of the original proof

In Theorem II of [1], it was stated that the lattice factorization in (5) is complete, i.e., it covers all possible solutions

of even-channel LPPRFBs with filter length of  $L = KM$  each. However, in the following, we shall show that this assertion is incorrect. To this end, let us briefly review the original incorrect proof in [1].

The proof of completeness relies on the existence of a building block  $\mathbf{G}_{K-1}(z)$  that can reduce the order of  $\mathbf{E}(z)$  by 1 at a time while retaining the LPPR property of the reduced-order  $\mathbf{F}(z) = \mathbf{G}_{K-1}^{-1}(z)\mathbf{E}(z)$ . A critical step in the proof is the *causality* of the factorization, i.e., there always exist invertible matrices  $\mathbf{U}_{K-1}$  and  $\mathbf{V}_{K-1}$  that can produce a causal  $\mathbf{F}(z)$ . Let  $\mathbf{F}(z) = \sum_{i=0}^{K-2} \mathbf{F}_i z^{-i}$  ( $\mathbf{F}_{K-2} \neq 0$ ) and  $\mathbf{E}(z) = \sum_{i=0}^{K-1} \mathbf{E}_i z^{-i}$  ( $\mathbf{E}_{K-1} \neq 0$ ), the proof of causality is equivalent to showing that there always exist two invertible matrices  $\mathbf{U}_{K-1}$  and  $\mathbf{V}_{K-1}$  such that ((A.4), [1])

$$\begin{bmatrix} \mathbf{U}_{K-1}^{-1} & -\mathbf{V}_{K-1}^{-1} \\ -\mathbf{U}_{K-1}^{-1} & \mathbf{V}_{K-1}^{-1} \end{bmatrix} \cdot \mathbf{E}_0 = \mathbf{0}_M. \quad (9)$$

In [1], it is presumed that (9) is satisfied if  $\text{rank}(\mathbf{E}_0) \leq M/2$ , since, in that case, the dimension of the null space of  $\mathbf{E}_0$  is larger than or equal to  $M/2$ , i.e.,  $\text{rank}(\text{Null}(\mathbf{E}_0)) \geq M/2$ , in which  $\text{Null}(\mathbf{E}_0)$  denotes the null space of  $\mathbf{E}_0$ . It is assumed in [1] that *under this condition, it is possible to choose  $M/2$  linearly independent vectors from  $\mathbf{E}_0$ 's null space to serve as  $[\mathbf{U}_{K-1}^{-1} \quad -\mathbf{V}_{K-1}^{-1}]$ .*

As previously mentioned,  $\mathbf{G}_i(z)$  in (6) can be replaced by  $\mathbf{G}_i(z)$  in (8) without loss of generality. Thus, we can rephrase the above assumption in the simplified factorization as follows. If  $\text{rank}(\text{Null}(\mathbf{E}_0)) \geq M/2$ , there always exist an invertible matrix  $\mathbf{U}_{K-1}$  such that

$$\begin{bmatrix} \mathbf{U}_{K-1}^{-1} & -\mathbf{I}_m \\ -\mathbf{U}_{K-1}^{-1} & \mathbf{I}_m \end{bmatrix} \cdot \mathbf{E}_0 = \mathbf{0}_M. \quad (10)$$

However, such assumption is *not* true.

Now, express  $\mathbf{E}_0$  as

$$\mathbf{E}_0 = \begin{bmatrix} \mathbf{E}_u \\ \mathbf{E}_d \end{bmatrix}, \quad (11)$$

where  $\mathbf{E}_u$  and  $\mathbf{E}_d$  are the upper and lower submatrices of  $\mathbf{E}_0$  with size of  $M/2 \times M$  each. Substituting (11) into (10) yields

$$\mathbf{E}_u = \mathbf{U}_{K-1} \mathbf{E}_d, \quad (12)$$

which indicates that  $\mathbf{E}_u$  can be represented through a linear transform of  $\mathbf{E}_d$ . In other words, (10) is satisfied *if and only if* the row vectors of  $\mathbf{E}_u$  and  $\mathbf{E}_d$  span the same space, *the condition  $\text{rank}(\text{Null}(\mathbf{E}_0)) \geq M/2$  alone cannot guarantee the existence of  $\mathbf{U}_{K-1}$  in (10)*. Hence, the original proof of completeness in [1] is incorrect and consequently, the completeness of the lattice factorization needs to be re-studied. In the following, through counter examples, we will show that the factorization is *not* complete when  $K > 2$ . Then, we prove that the factorization is complete when  $K \leq 2$ .

## B. $K > 2$

Actually, (12) is a very strong restriction. When  $K > 2$ , an LPPRFB can exist even if (12) does not hold. To see this more clearly, we first provide a counter example for  $K = 3$ . Based on it, other counter examples can be generated for  $K > 3$ .

### B.1 $K = 3$

A counter example for  $K = 3$  is as follows. For instance, when  $M = 4$ , let  $\mathbf{E}(z)$  and  $\mathbf{R}(z)$  be chosen as

$$\begin{aligned} \mathbf{E}(z) &= \begin{bmatrix} 1 & 1 & 1 & 1 \\ -1 & -1 & -1 & -1 \\ 1 & -1 & 1 & -1 \\ -1 & 1 & -1 & 1 \end{bmatrix} + \begin{bmatrix} 2 & 0 & 0 & 2 \\ 0 & 2 & 2 & 0 \\ 2 & 0 & 0 & -2 \\ 0 & -2 & 2 & 0 \end{bmatrix} z^{-1} \\ &+ \begin{bmatrix} 1 & 1 & 1 & 1 \\ -1 & -1 & -1 & -1 \\ 1 & -1 & 1 & -1 \\ -1 & 1 & -1 & 1 \end{bmatrix} z^{-2}, \end{aligned} \quad (13)$$

and

$$\begin{aligned} \mathbf{R}(z) &= \begin{bmatrix} -\frac{1}{8} & -\frac{1}{8} & -\frac{1}{8} & -\frac{1}{8} \\ \frac{1}{8} & \frac{1}{8} & \frac{1}{8} & \frac{1}{8} \\ -\frac{1}{8} & -\frac{1}{8} & -\frac{1}{8} & -\frac{1}{8} \\ \frac{1}{8} & \frac{1}{8} & \frac{1}{8} & \frac{1}{8} \end{bmatrix} + \begin{bmatrix} \frac{1}{4} & 0 & \frac{1}{4} & 0 \\ 0 & \frac{1}{4} & 0 & -\frac{1}{4} \\ 0 & \frac{1}{4} & 0 & \frac{1}{4} \\ \frac{1}{4} & 0 & -\frac{1}{4} & 0 \end{bmatrix} z^{-1} \\ &+ \begin{bmatrix} -\frac{1}{8} & -\frac{1}{8} & -\frac{1}{8} & -\frac{1}{8} \\ \frac{1}{8} & \frac{1}{8} & \frac{1}{8} & \frac{1}{8} \\ -\frac{1}{8} & -\frac{1}{8} & -\frac{1}{8} & -\frac{1}{8} \\ \frac{1}{8} & \frac{1}{8} & \frac{1}{8} & \frac{1}{8} \end{bmatrix} z^{-2}. \end{aligned} \quad (14)$$

One can verify that  $\mathbf{E}(z)$  and  $\mathbf{R}(z)$  satisfy

$$\mathbf{R}(z)\mathbf{E}(z) = z^{-2}\mathbf{I}_4, \quad (15)$$

$$\mathbf{E}(z) = z^{-2}\mathbf{D}_4\mathbf{E}(z^{-1})\mathbf{J}_4, \quad (16)$$

$$\mathbf{R}(z) = z^{-2}\mathbf{J}_4\mathbf{R}(z^{-1})\mathbf{D}_4. \quad (17)$$

That is,  $\mathbf{E}(z)$  and  $\mathbf{R}(z)$  satisfy the LPPR condition with  $l = 2$ ,  $K = 3$  and  $M = 4$  in (2)-(4). Moreover, from (13), it is easy to see that  $\text{rank}(\mathbf{E}_0) = 4/2 = 2$ . However, in this example, the basis vector of  $\mathbf{E}_u$  is  $[1 \ 1 \ 1 \ 1]$ , while the basis vector of  $\mathbf{E}_d$  is  $[1 \ -1 \ 1 \ -1]$ . They do not span the same space. Thus,  $\mathbf{U}_2$  does not exist. As a result,  $\mathbf{E}(z)$  can not be expressed in terms of (5). This example means that the factorization cannot cover all the solutions when  $K = 3$ . Based on it, we can also find other counter examples for  $K > 3$  as follows.

### B.2 $K > 3$

When  $K$  is odd ( $K = 2K_0 + 1$ ,  $K_0 \geq 1$ ), let the corresponding analysis bank  $\mathbf{E}^o(z) = \sum_{i=0}^{K-1} \mathbf{E}_i^o z^{-i}$  and synthesis bank  $\mathbf{R}^o(z) = \sum_{i=0}^{K-1} \mathbf{R}_i^o z^{-i}$  be chosen as

$$\mathbf{E}^o(z) = \mathbf{E}(z^{K_0}) \quad \text{and} \quad \mathbf{R}^o(z) = \mathbf{R}(z^{K_0}), \quad (18)$$

in which  $\mathbf{E}(z)$  and  $\mathbf{R}(z)$  are the same as in (13) and (14), respectively. From (15)-(18), one can derive that

$$\begin{aligned} \mathbf{R}(z^{K_0})\mathbf{E}(z^{K_0}) &= z^{-2K_0}\mathbf{I}_4 \\ \implies \mathbf{R}^\circ(z)\mathbf{E}^\circ(z) &= z^{-2K_0}\mathbf{I}_4, \end{aligned} \quad (19)$$

$$\begin{aligned} \mathbf{E}(z^{K_0}) &= z^{-2K_0}\mathbf{D}_4\mathbf{E}(z^{-K_0})\mathbf{J}_4 \\ \implies \mathbf{E}^\circ(z) &= z^{-2K_0}\mathbf{D}_4\mathbf{E}^\circ(z^{-1})\mathbf{J}_4, \end{aligned} \quad (20)$$

$$\begin{aligned} \mathbf{R}(z) &= z^{-2K_0}\mathbf{J}_4\mathbf{R}(z^{-K_0})\mathbf{D}_4 \\ \implies \mathbf{R}^\circ(z) &= z^{-2K_0}\mathbf{J}_4\mathbf{R}^\circ(z^{-1})\mathbf{J}_4. \end{aligned} \quad (21)$$

In other words,  $\mathbf{E}^\circ(z)$  and  $\mathbf{R}^\circ(z)$  satisfy the LPPR condition in (2)-(4) with  $l = 2K_0$ ,  $K = 2K_0 + 1$  and  $M = 4$ . But since  $\mathbf{E}_0^\circ = \mathbf{E}_0$ , for the same reason as explained in the previous example,  $\mathbf{E}^\circ(z)$  can not be represented through (5), either.

On the other hand, when  $K$  is even ( $K = 2(K_0 + 1)$ ,  $K_0 \geq 1$ ), let the analysis bank  $\mathbf{E}^e(z) = \sum_{i=0}^{K-1} \mathbf{E}_i^e z^{-i}$  and synthesis bank  $\mathbf{R}^e(z) = \sum_{i=0}^{K-1} \mathbf{R}_i^e z^{-i}$  be chosen as

$$\begin{aligned} \mathbf{E}^e(z) &= \mathbf{E}^\circ(z) \text{diag}(\mathbf{I}_2, z^{-1}\mathbf{I}_2) \\ \mathbf{R}^e(z) &= \text{diag}(z^{-1}\mathbf{I}_2, \mathbf{I}_2)\mathbf{R}^\circ(z), \end{aligned} \quad (22)$$

where  $\mathbf{E}^\circ(z)$  and  $\mathbf{R}^\circ(z)$  are defined in (18) with  $K_0$  taking the same value. Since  $\mathbf{E}^\circ(z)$  and  $\mathbf{R}^\circ(z)$  satisfy the PR property in (2), it is easy to verify that  $\mathbf{E}^e(z)$  and  $\mathbf{R}^e(z)$  also satisfy the PR property as follows.

$$\begin{aligned} \mathbf{R}^e(z)\mathbf{E}^e(z) &= \text{diag}(z^{-1}\mathbf{I}_2, \mathbf{I}_2)\mathbf{R}^\circ(z)\mathbf{E}^\circ(z)\text{diag}(\mathbf{I}_2, z^{-1}\mathbf{I}_2) \\ &= z^{-(2K_0+1)}\mathbf{I}_4. \end{aligned} \quad (23)$$

Moreover, noticing the fact that

$$\text{diag}(\mathbf{I}_2, z^{-1}\mathbf{I}_2) = z^{-1}\mathbf{J}_4\text{diag}(\mathbf{I}_2, z\mathbf{I}_2)\mathbf{J}_4, \quad (24)$$

and using (20), we can derive that  $\mathbf{E}^e(z)$  satisfies the LP property as follows

$$\begin{aligned} \mathbf{E}^e(z) &= \mathbf{E}^\circ(z)\text{diag}(\mathbf{I}_2, z^{-1}\mathbf{I}_2) \\ &= z^{-2K_0}\mathbf{D}_4\mathbf{E}^\circ(z^{-1})\mathbf{J}_4 z^{-1}\mathbf{J}_4\text{diag}(\mathbf{I}_2, z\mathbf{I}_2)\mathbf{J}_4 \\ &= z^{-(2K_0+1)}\mathbf{D}_4\mathbf{E}^e(z^{-1})\mathbf{J}_4 \\ &= z^{-(K-1)}\mathbf{D}_4\mathbf{E}^e(z^{-1})\mathbf{J}_4. \end{aligned} \quad (25)$$

In a similar way, one can verify that  $\mathbf{R}^e(z)$  satisfies the LP property also

$$\begin{aligned} \mathbf{R}^e(z) &= z^{-(2K_0+1)}\mathbf{J}_4\mathbf{R}^e(z^{-1})\mathbf{D}_4 \\ &= z^{-(K-1)}\mathbf{J}_4\mathbf{R}^e(z^{-1})\mathbf{D}_4. \end{aligned} \quad (26)$$

Equations (23) and (25)-(26) mean that  $\mathbf{E}^e(z)$  and  $\mathbf{R}^e(z)$  satisfy the LPPR condition in (2)-(4) with  $l = 2K_0 + 1$ ,  $K = 2(K_0 + 1)$  and  $M = 4$ . However, from (22), one can compute that

$$\mathbf{E}_0^e = \mathbf{E}_0^\circ \begin{bmatrix} \mathbf{I}_2 & \mathbf{0}_2 \\ \mathbf{0}_2 & \mathbf{0}_2 \end{bmatrix} = \mathbf{E}_0^\circ \begin{bmatrix} \mathbf{I}_2 & \mathbf{0}_2 \\ \mathbf{0}_2 & \mathbf{0}_2 \end{bmatrix} = \begin{bmatrix} 1 & 1 & 0 & 0 \\ -1 & -1 & 0 & 0 \\ 1 & -1 & 0 & 0 \\ -1 & 1 & 0 & 0 \end{bmatrix}. \quad (27)$$

Let  $\mathbf{E}_u^e$  and  $\mathbf{E}_d^e$  denote the first two rows and the last two rows of  $\mathbf{E}_0^e$ , respectively, i.e.,

$$\mathbf{E}_u^e = \begin{bmatrix} 1 & 1 & 0 & 0 \\ -1 & -1 & 0 & 0 \end{bmatrix} \quad \text{and} \quad \mathbf{E}_d^e = \begin{bmatrix} 1 & -1 & 0 & 0 \\ -1 & 1 & 0 & 0 \end{bmatrix}. \quad (28)$$

Clearly, the row vectors of  $\mathbf{E}_u^e$  and  $\mathbf{E}_d^e$  do not span the same space. Therefore,  $\mathbf{E}^e(z)$  cannot be represented in terms of (5), either.

From the above counter examples, we can conclude that the lattice factorization in (5) is *not* complete for  $K > 2$ .

### C. $K \leq 2$

Now, let us study the case when  $K \leq 2$ . It is easy to see that when  $K = 1$ , the factorization is complete [1]. Next, we will prove that the factorization is also complete when  $K = 2$ .

Similarly as in [1], we have to show the existence of a  $\mathbf{G}_1^{-1}(z)$  such that  $\mathbf{F}(z) = \mathbf{G}_1^{-1}(z)\mathbf{E}(z)$  corresponds to an LPPRFB with  $K = 1$ . Since the preservation of LPPR property and order reduction property have been verified in [1], we need only to prove causality. The proof of completeness for  $K = 2$  is accomplished if we can verify the existence of an invertible matrix  $\mathbf{U}_1$  satisfying (10), or equivalently, (12). Just as in [1], for simplicity of exposition, denote  $\mathbf{E}(z) = \mathbf{E}_0 + \mathbf{E}_1 z^{-1}$  and  $\mathbf{R}(z) = \mathbf{R}_0 + \mathbf{R}_1 z$ . The LP condition implies that [1]

$$\mathbf{E}_1 = \mathbf{D}_M \mathbf{E}_0 \mathbf{J}_M \quad \text{and} \quad \mathbf{R}_1 = \mathbf{J}_M \mathbf{R}_0 \mathbf{D}_M. \quad (29)$$

Equivalently,  $\mathbf{E}_i$  and  $\mathbf{R}_i$  (for  $i = 0, 1$ ) should take the following forms

$$\mathbf{E}_0 = \begin{bmatrix} \mathbf{E}_u \\ \mathbf{E}_d \end{bmatrix}, \quad \mathbf{E}_1 = \begin{bmatrix} \mathbf{E}_u \mathbf{J}_M \\ -\mathbf{E}_d \mathbf{J}_M \end{bmatrix}; \quad (30)$$

and

$$\mathbf{R}_0 = [\mathbf{R}_l \quad \mathbf{R}_r], \quad \mathbf{R}_1 = [\mathbf{J}_M \mathbf{R}_l \quad -\mathbf{J}_M \mathbf{R}_r]; \quad (31)$$

in which  $\mathbf{E}_u$  and  $\mathbf{E}_d$  are, respectively, the upper and lower submatrices of  $\mathbf{E}_0$  with size of  $M/2 \times M$  each, while  $\mathbf{R}_l$  and  $\mathbf{R}_r$  are, respectively, the left and right submatrices of  $\mathbf{R}_0$  with size of  $M \times M/2$  each. Besides, the PR property  $\mathbf{R}(z)\mathbf{E}(z) = \mathbf{I}_M$  leads to

$$\mathbf{E}(z)\mathbf{R}(z) = \mathbf{I}_M, \quad (32)$$

from which one can derive that

$$\mathbf{E}_0 \mathbf{R}_0 + \mathbf{E}_1 \mathbf{R}_1 = \mathbf{I}_M, \quad (33)$$

and

$$\mathbf{E}_0 \mathbf{R}_1 = \mathbf{0}_M. \quad (34)$$

With (29), one can obtain

$$\mathbf{E}_0 \mathbf{R}_1 = \mathbf{E}_0 \mathbf{J}_M \mathbf{R}_0 \mathbf{D}_M = \mathbf{0}_M \implies \mathbf{E}_0 \mathbf{J}_M \mathbf{R}_0 = \mathbf{0}_M. \quad (35)$$

Substituting (30) and (31) into (33) yields

$$2\mathbf{E}_u\mathbf{R}_l = 2\mathbf{E}_d\mathbf{R}_r = \mathbf{I}_{M/2}. \quad (36)$$

As for an  $x \times y$  matrix  $\mathbf{A}$  and a  $y \times z$  matrix  $\mathbf{B}$ ,  $\min(\text{rank}(\mathbf{A}), \text{rank}(\mathbf{B})) \geq \text{rank}(\mathbf{AB})$ , thus,

$$\begin{aligned} \text{rank}(\mathbf{E}_u) &\geq M/2, & \text{rank}(\mathbf{E}_d) &\geq M/2, \\ \text{rank}(\mathbf{R}_l) &\geq M/2, & \text{rank}(\mathbf{R}_r) &\geq M/2. \end{aligned} \quad (37)$$

Besides, for an  $x \times y$  matrix  $\mathbf{A}$ ,  $\text{rank}(\mathbf{A}) \leq \min(\mathbf{x}, \mathbf{y})$ , hence,

$$\begin{aligned} \text{rank}(\mathbf{E}_u) &\leq M/2, & \text{rank}(\mathbf{E}_d) &\leq M/2, \\ \text{rank}(\mathbf{R}_l) &\leq M/2, & \text{rank}(\mathbf{R}_r) &\leq M/2. \end{aligned} \quad (38)$$

Based on (37) and (38), we can deduce that

$$\begin{aligned} \text{rank}(\mathbf{E}_u) &= \text{rank}(\mathbf{E}_d) \\ = \text{rank}(\mathbf{R}_l) &= \text{rank}(\mathbf{R}_r) = M/2, \end{aligned} \quad (39)$$

which means that all the  $M/2$  rows of  $\mathbf{E}_u$  and  $\mathbf{E}_d$  must be independent, and all the  $M/2$  columns of  $\mathbf{R}_l$  and  $\mathbf{R}_r$  are independent, too. Hence,

$$\text{rank}(\mathbf{E}_0) \geq M/2 \quad \text{and} \quad \text{rank}(\mathbf{R}_0) \geq M/2. \quad (40)$$

Furthermore, applying Sylvester's rank inequality [3] to (35) results in

$$\begin{aligned} \text{rank}(\mathbf{E}_0) + \text{rank}(\mathbf{R}_0) - M &\leq \text{rank}(\mathbf{E}_0\mathbf{J}_M\mathbf{R}_0) = 0 \\ \implies \text{rank}(\mathbf{E}_0) + \text{rank}(\mathbf{R}_0) &\leq M. \end{aligned} \quad (41)$$

From (40) and (41), we have

$$\text{rank}(\mathbf{E}_0) = \text{rank}(\mathbf{R}_0) = M/2. \quad (42)$$

Together, (39) and (42) mean that the row vectors of  $\mathbf{E}_u$  and  $\mathbf{E}_d$  must span the same space. Hence,  $\mathbf{E}_u$  can be represented through a linear transform of  $\mathbf{E}_d$ . That is, there always exists an invertible matrix  $\mathbf{U}_1$  such that (12), or equivalently, (10) exists, which further means that causality is met.

From the above discussions, Theorem II in [1] should be amended as follows.

*Theorem 1:* For any  $M$ -channel ( $M$  even) LPPRFB with all the analysis and synthesis filters of length  $L = 2M$  each, the corresponding analysis polyphase matrix  $\mathbf{E}(z)$  can always be factored as in (5).

#### IV. CONCLUSION

We have re-examined the completeness of the lattice factorization for  $M$ -channel ( $M$  even) LPPRFB with all the filters of the same length  $L = KM$  in [1]. We point out that the original proof of completeness contains an incorrect assumption. Through counter examples, we show that the factorization is not complete for  $K > 2$ . Additionally, we prove that the factorization is complete when  $K \leq 2$ . The complete factorization of the general  $L = KM$  ( $K > 2$ ) case is still an open problem.

#### REFERENCES

- [1] T.D. Tran, R.L. de Queiroz, and T.Q. Nguyen, "Linear-phase perfect reconstruction filter bank: lattice structure, design, and application in image coding," *IEEE Trans. Signal Processing*, vol. 48, pp. 133-147, Jan. 2000.
- [2] L. Gan and K.-K. Ma, "A simplified lattice factorization for linear-phase perfect reconstruction filter bank," *IEEE Signal Processing Lett.*, vol. 8, pp. 207-209, Jul. 2001.
- [3] F.R. Gantmacher, *The Theory of Matrices*, New York: Chelsea, 1977.